



FUNDAMENTALS OF ELECTRIC CIRCUITS

Part 1: DC CIRCUITS



Chapter 8: Second-order Circuits

I. Introduction.

II. Finding initial and final values.

III. The source free series/parallel RLC circuits.

IV. Step response of a series/parallel RLC circuits.

V. General second-order circuits.

VI. Applications.

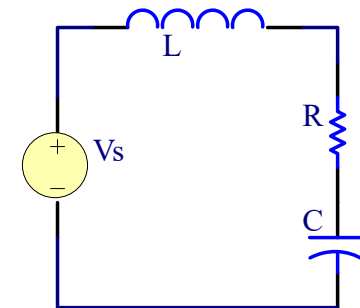


Chapter 8: Second-order circuits



I. Introduction

- In the previous chapter, we considered circuits with a single storage element (C or L). Such circuits are first-order because the differential equations describing them are first-order.
- In this chapter, we will consider circuits containing two storage elements (known as *second-order circuits*).
- A *second-order circuit* is characterized by a second-order differential equation. It consists of resistors and the equivalent of 02 energy storage elements.
- There are three kinds of second-order circuits:
 - ❖ Two storage elements of different type: L and C
 - ❖ Two storage elements of one type: L or C
 - ❖ An op amp circuit with 02 storage elements



Series RLC circuit

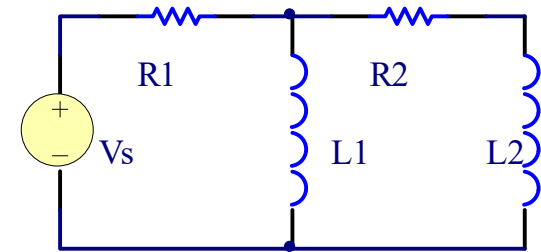
Chapter 8: Second-order circuits

I. Introduction

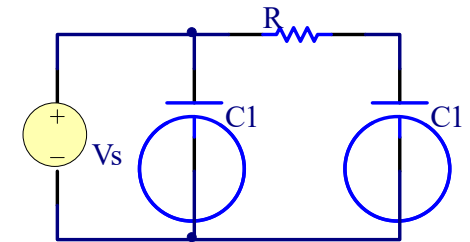
➤ The analysis of second-order will be similar to that used for first-order:

❖ First, consider circuits that are excited by the initial conditions of the storage elements → source free circuitss give natural responses.

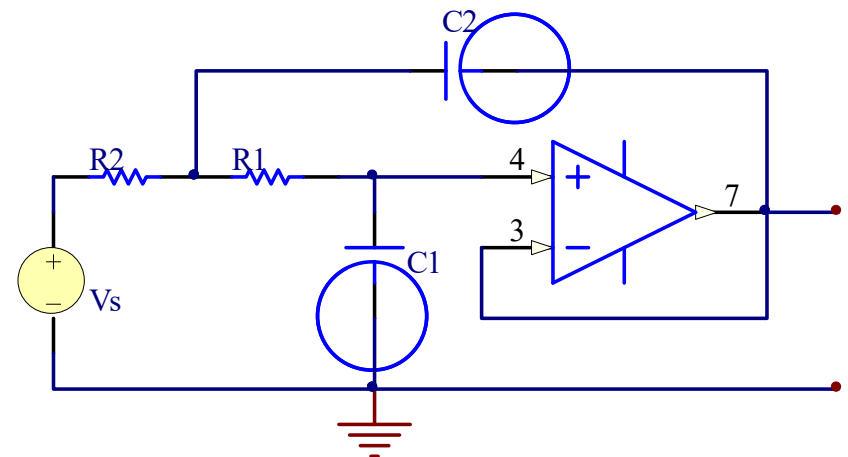
❖ Second, with independent sources, circuits will give both the nature response and the forced response.



RL circuit



RC circuit



Op amp with 2 storage element



Chapter 8: Second-order circuits



II. Finding initial and final values

- The major problem in handling second-order circuits is finding the initial and final conditions on circuits variables:
 - ❖ Easy to get the initial and final values of v and i
 - ❖ Difficult to find the initial values of their derivatives: dv/dt , di/dt
- 02 key points in determining the initial conditions:
 - ❖ First, carefully handle the *polarity of voltage $v_C(t)$* , and the *directions of $i_L(t)$*
 - ❖ Second, keep in mind that *$v_C(t)$, $i_L(t)$ are always continuous*:

$$V_C(+0) = V_C(-0) \quad ; \quad i_L(+0) = i_L(-0)$$

where:

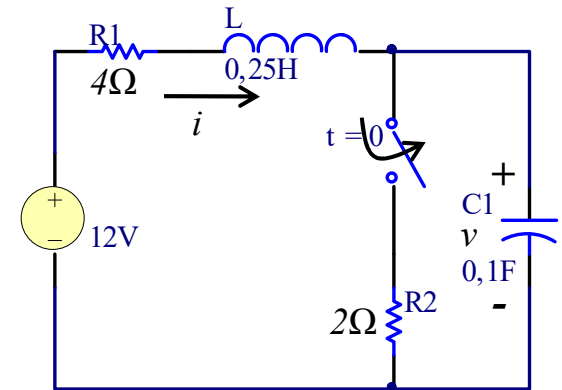
$t = 0$: the time that the switching event takes place

$t = -0$: the time just before a switching event

$t = +0$: the time just after a switching event

II. Finding initial and final values

Ex 8.1: The switch has been closed for a long time, and opens at $t = 0$. Find $i(+0)$, $v(+0)$, $di(+0)/dt$, $dv(+0)/dt$, $i(\infty)$, $v(\infty)$

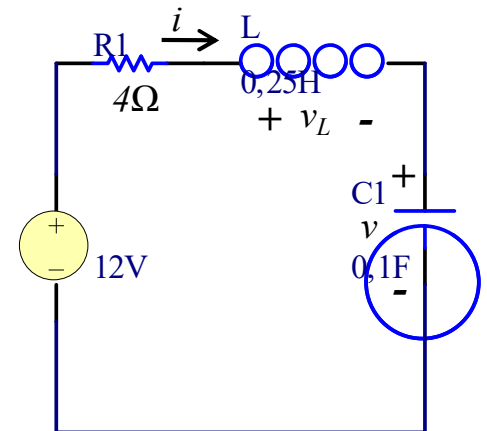


- $t = -0$: The circuit has reached DC steady state, so L acts like a short circuit, C acts like an open circuit

$$i(-0) = \frac{12}{R_1 + R_2} = 2A, \quad v(-0) = 2i(-0) = 4V$$

- As $v_C(t)$ and $i_L(t)$ cannot change abruptly

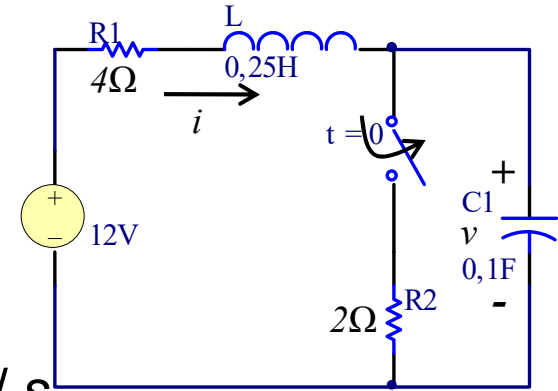
$$v_C(+0) = v_C(-0) = 4V ; \quad i_L(+0) = i_L(-0) = 2A$$



Chapter 8: Second-order circuits

II. Finding initial and final values

Ex 8.1: The switch has been closed for a long time, and opens at $t = 0$. Find $i(+0)$, $v(+0)$, $di(+0)/dt$, $dv(+0)/dt$, $i(\infty)$, $v(\infty)$

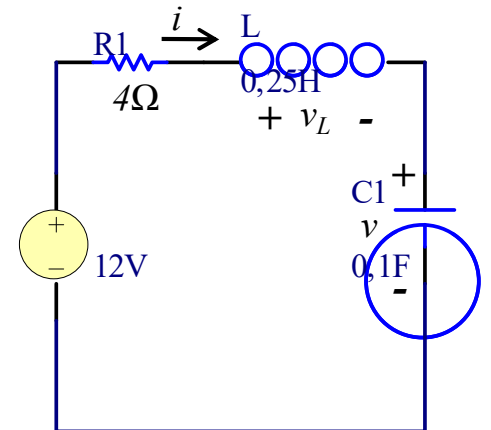


$$\text{➤ } t = +0: i_C = C \frac{dv_C}{dt} \rightarrow \frac{dv_C(+0)}{dt} = \frac{i_C(+0)}{C} = \frac{2}{0,1} = 20V/s$$

➤ Applying KVL:

$$-12 + 4i(+0) + v_L(+0) + v_C(+0) = 0 \rightarrow v_L(+0) = 0$$

$$\rightarrow \frac{di(+0)}{dt} = \frac{v_L(+0)}{L} = 0A/s$$



➤ $t > 0$: the circuit undergoes transience

➤ $t > \infty$: Circuit reaches steady state again: $i(\infty) = 0A$, $v(\infty) = 12V$

II. Finding initial and final values

Ex 8.2: Find $i_L(+0)$, $v_C(+0)$, $v_R(+0)$, $di_L(+0)/dt$, $dv_C(+0)/dt$, $i_L(\infty)$, $v_C(\infty)$, $v_R(\infty)$

- For $t = < 0$: $3.u(t) = 0$
- At $t = -0$: Circuit has reached steady state

$$i_L(-0) = 0, v_R(-0) = 0, v_C(-0) = -20V$$

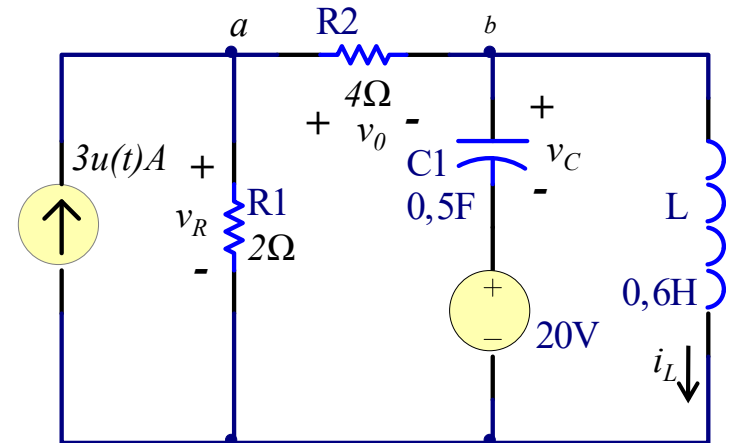
- For $t > 0$: $3.u(t) = 3$. Since i_L and u_C cannot change abruptly

$$i_L(+0) = i_L(-0) = 0, v_C(+0) = v_C(-0) = -20V$$

- Apply KCL at node a : $3 = \frac{v_R(+0)}{R_1} + \frac{v_0(+0)}{R_2} \rightarrow v_R(+0) = v_0(+0) = 4V$

- Apply KVL to the middle mesh:

$$-v_R(+0) + v_0(+0) + v_C(+0) + 20 = 0 \xrightarrow{v_C(+0) = -20V} v_R(+0) = v_0(+0)$$



II. Finding initial and final values

Ex 8.2: Find $i_L(+0)$, $v_C(+0)$, $v_R(+0)$, $di_L(+0)/dt$, $dv_C(+0)/dt$, $i_L(\infty)$, $v_C(\infty)$, $v_R(\infty)$

➤ Since $L \cdot di_L/dt = v_L \rightarrow \frac{di_L(+0)}{dt} = \frac{v_L(+0)}{L}$

➤ Applying KVL in the right mesh gives:

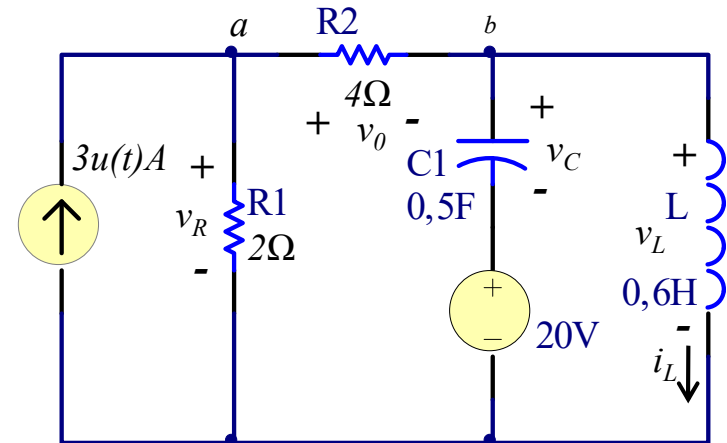
$$v_L(+0) = v_C(+0) + 20 = -20 + 20 = 0 \rightarrow \frac{di_L(+0)}{dt} = 0$$

➤ Apply KCL at node b : $\frac{v_0(+0)}{R_2} = i_C(+0) + i_L(+0) \xrightarrow[i_L(+0)=0]{v_0(+0)=4} i_C(+0) = 1A$

$$\rightarrow \frac{dv_C(+0)}{dt} = \frac{i_C(+0)}{C} = \frac{1}{0,5} = 2V/s$$

➤ Apply KCL to node a , and taking the derivative of each term and setting $t = +0$:

$$3 = \frac{v_R}{R_1} + \frac{v_0}{R_2} \leftrightarrow 0 = 2 \frac{dv_R(+0)}{dt} + \frac{dv_0(+0)}{dt}$$





Chapter 8: Second-order circuits

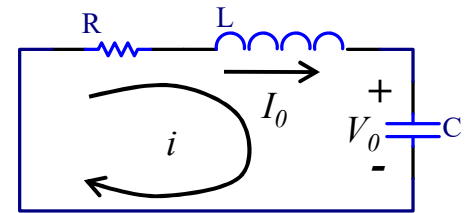


III. The source-free series/parallel *RLC* circuit

III.1. The source-free series *RLC* circuit

- An understanding of the natural response of the series *RLC* circuit is a necessary background for future studies in *filter design* and *communications networks*.

- Consider the series *RLC* circuit that is excited by the energy initially stored V_0 , and I_0 .



- At $t = 0$: $v(0) = \frac{1}{C} \int_{-\infty}^0 i dt = V_0$, $i(0) = I_0$
- Applying KVL to the loop: $Ri + L \frac{di}{dt} + \frac{1}{C} \int_{-\infty}^t i dt = 0 \leftrightarrow \frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = 0$
- To solve a second-order differential equation it requires 02 initial conditions:
 - ❖ $i(+0)$ and $i'(+0)$, or
 - ❖ $i(+0)$ and $v(+0)$

III. The source-free series/parallel *RLC* circuit

III.1. The source-free series *RLC* circuit

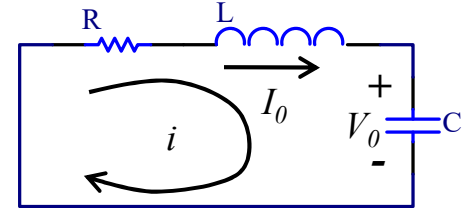
➤ First, we find $i'(0)$: $Ri(0) + L \frac{di(0)}{dt} + V_0 = 0 \rightarrow \frac{di(0)}{dt} = -\frac{1}{L}(Ri_0 + V_0)$

➤ The solution of the second-order equation is: $i = A.e^{st}$

➤ Substituting the solution into the equation has:

$$As^2 e^{st} + \frac{AR}{L} s e^{st} + \frac{A}{LC} e^{st} = 0$$

$$\rightarrow A e^{st} \underbrace{\left(s^2 + \frac{R}{L} s + \frac{1}{LC} \right)}_{\text{characteristic equation}} = 0 \rightarrow \begin{cases} s_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}} = -\alpha + \sqrt{\alpha^2 - \omega_0^2} & \alpha = \frac{R}{2L} \\ s_2 = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}} = -\alpha - \sqrt{\alpha^2 - \omega_0^2} & \omega_0 = \frac{1}{\sqrt{LC}} \end{cases}$$



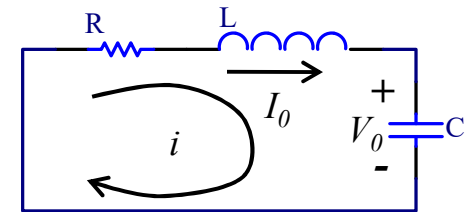
❖ roots s_1 and s_2 are called *natural frequencies* [Np/s]

❖ ω_0 : *resonant frequency* (un-damped natural frequency, or damping factor)

III. The source-free series/parallel *RLC* circuit

III.1. The source-free series *RLC* circuit

$$\begin{cases} s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2} \\ s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2} \end{cases} \quad \alpha = \frac{R}{2L}, \quad \omega_0 = \frac{1}{\sqrt{LC}}$$



➤ There are three types of solutions:

- ❖ If $\alpha > \omega_0$: over damped case
- ❖ If $\alpha = \omega_0$: critically damped case
- ❖ If $\alpha < \omega_0$: under damped case

➤ 02 values of s show that there are 02 possible solutions for i

$$i_1 = A_1 \cdot e^{s_1 t}, \quad i_2 = A_2 \cdot e^{s_2 t} \rightarrow i(t) = A_1 \cdot e^{s_1 t} + A_2 \cdot e^{s_2 t}$$

A_1, A_2 are determined from the initial values $i(0)$ and $di(0)/dt$.

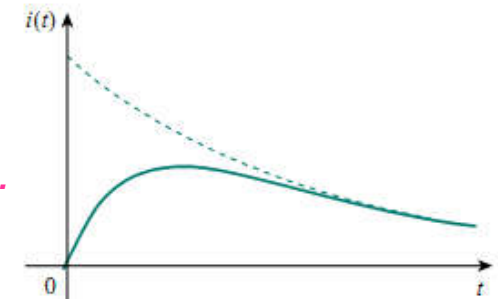
III. The source-free series/parallel *RLC* circuit

III.1. The source-free series *RLC* circuit

- *Over damped* case $\alpha > \omega_0 \rightarrow s_1$ and s_2 are *negative* and *real*.

$$i(t) = A_1 \cdot e^{s_1 t} + A_2 \cdot e^{s_2 t}$$

Response *decays* and *approaches* zero as t increases.

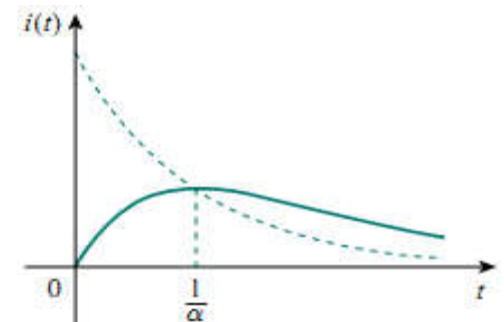


A typical over damped response

- *Critically damped* case $\alpha = \omega_0$

$$i(t) = (A_1 \cdot t + A_2) \cdot e^{-\alpha t}$$

Response reaches a maximum value at $t = 1/\alpha$, and then *decays* all the way to zero.



A typical critically damped response

III. The source-free series/parallel *RLC* circuit

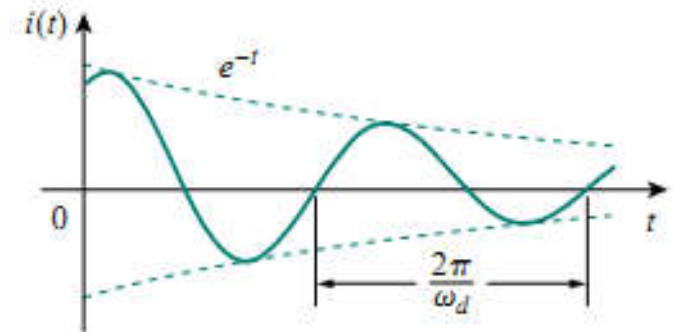
III.1. The source-free series *RLC* circuit

➤ *Under damped* case $\alpha < \omega_0$:

$$\rightarrow \begin{cases} s_1 = -\alpha + \sqrt{-(\omega_0^2 - \alpha^2)} = -\alpha + j\omega_d \\ s_2 = -\alpha - \sqrt{-(\omega_0^2 - \alpha^2)} = -\alpha - j\omega_d \end{cases} \quad \begin{array}{l} \omega_d: \text{damped natural frequency} \\ \omega_0: \text{un-damped natural frequencies} \end{array}$$

The natural response is:

$$i(t) = e^{-\alpha t} (B_1 \cos \omega_d t + B_2 \sin \omega_d t)$$



A typical under damped response

- ❖ The natural response is *exponentially damped* and *oscillatory* in nature.
- ❖ The response has a time constant of $1/\alpha$ and a period of $T = 2\pi/\omega_d$



Chapter 8: Second-order circuits



III. The source-free series/parallel *RLC* circuit

III.1. The source-free series *RLC* circuit

➤ Notes:

- ❖ *Damping effect* is due to the presence of R
- ❖ *Damping factor* α determines the rate at which the response is damped:
 - ❑ $\alpha = 0$: having LC circuit with $1/\sqrt{LC}$ as the un-damped natural frequency
 - ❑ $\alpha < \omega_0$: response is not only damped but also oscillatory.
- ❖ By adjusting R , response may be made un-damped, over damped, critically damped, or under damped.
- ❖ *Oscillatory response* is possible due to the presence of L , C : They allow the flow of energy back and forth.
- ❖ The *critically damped* case is the *borderline* between the under damped and over damped cases.



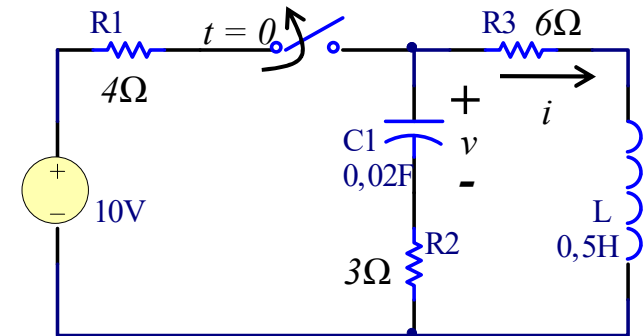
Chapter 8: Second-order circuits



III. The source-free series/parallel *RLC* circuit

III.1. The source-free series *RLC* circuit

Ex 8.3: Find $i(t)$ in the circuit. Assume that the circuit has reached steady state at $t = -0$.



➤ For $t < 0$: $i(0) = \frac{10}{R_1 + R_3} = 1A$; $v(0) = i \cdot R_3 = 6V$

➤ For $t > 0$: Source – free series *RLC* circuit

❖ The roots:

$$\alpha = \frac{R_{eq}}{2L} = \frac{9}{2 \cdot 0,5} = 9, \quad \omega_0 = \frac{1}{\sqrt{LC}} = 10, \quad s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} = -9 \pm j4,359$$

❖ The response is under damped: $i(t) = e^{-9t} (A_1 \cos 4,359t + A_2 \sin 4,359t)$

III. The source-free series/parallel RLC circuit

III.2. The source-free parallel RLC circuit

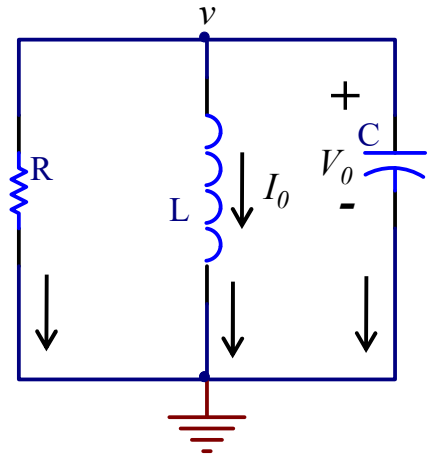
➤ Applying KCL: $\frac{v}{R} + \frac{1}{L} \int_{-\infty}^t v dt + C \frac{dv}{dt} = 0 \rightarrow \frac{d^2 v}{dt^2} + \frac{1}{RC} \frac{dv}{dt} + \frac{1}{LC} v = 0$

$$s^2 + \frac{1}{RC} s + \frac{1}{LC} = 0$$

$$\rightarrow s_{1,2} = -\frac{1}{2RC} \pm \sqrt{\left(\frac{1}{2RC}\right)^2 - \frac{1}{LC}} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2}$$

$$\alpha = \frac{1}{2RC}$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$



❖ Over damped ($\alpha > \omega_0$): $v(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$

❖ Critically damped ($\alpha = \omega_0$): $v(t) = (A_1 + A_2 t) e^{-\alpha t}$

❖ Under damped ($\alpha < \omega_0$): $s_{1,2} = -\alpha \pm j\omega_d$, $\omega_d = \sqrt{\omega_0^2 - \alpha^2}$

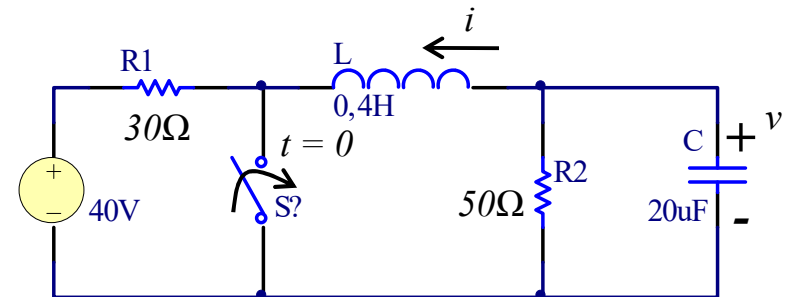
$$v(t) = e^{-\alpha t} (A_1 \cos \omega_d t + A_2 \sin \omega_d t)$$

A_1, A_2 : determine from the initial conditions: $v(0)$, $dv(0)/dt$

III. The source-free series/parallel *RLC* circuit

III.2. The source-free parallel *RLC* circuit

Ex 8.4: Find $v(t)$ for $t > 0$ in the RLC circuit



- When $t < 0$: The switch is opened.

$$v(0) = \frac{R_2}{R_1 + R_2} 40 = 25V, \quad i(0) = -\frac{40}{R_1 + R_2} = -0,5A$$

- At $t = 0$: Applying the KCL gives:

$$\frac{v(0)}{R} + \frac{1}{L} \int_{-\infty}^0 v dt + C \frac{dv(0)}{dt} = 0 \rightarrow \frac{v(0)}{R} + i(0) + C \frac{dv(0)}{dt} = 0$$

$$\rightarrow \frac{dv(0)}{dt} = -\frac{v(0) + Ri(0)}{RC} = -\frac{25 - 50 \cdot 0,5}{50 \cdot 20 \cdot 10^{-6}} = 0$$



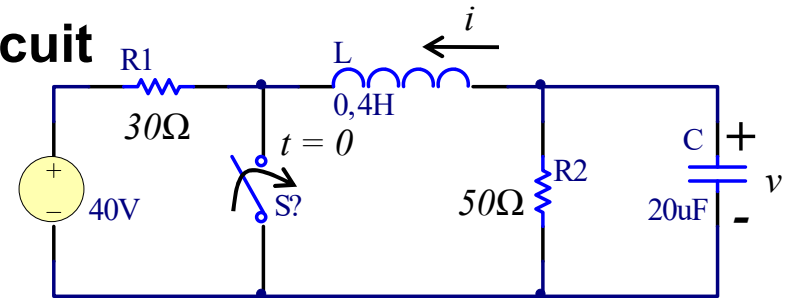
Chapter 8: Second-order circuits



III. The source-free series/parallel *RLC* circuit

III.2. The source-free parallel *RLC* circuit

Ex 8.4: Find $v(t)$ for $t > 0$ in the RLC circuit



➤ When $t > 0$: The switch is closed. $\alpha = \frac{1}{2RC} = 500$, $\omega_0 = \frac{1}{\sqrt{LC}} = 354$

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} \rightarrow \begin{cases} s_1 = -854 \\ s_2 = -146 \end{cases} \rightarrow v(t) = A_1 e^{-854t} + A_2 e^{-146t}$$

$$\text{At } t = 0: \begin{cases} v(0) = A_1 + A_2 = 25 \\ \frac{dv(0)}{dt} = -854A_1 - 146A_2 = 0 \end{cases} \rightarrow \begin{cases} A_1 = -5,16 \\ A_2 = 30,16 \end{cases}$$

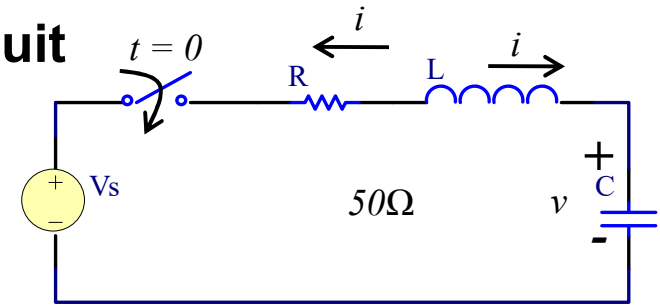
➤ The complete solution:

$$v(t) = -5,16e^{-854t} + 30,16e^{-146t} \text{ V}$$

IV. Step response of a series/parallel *RLC* circuit

IV.1. Step response of a series *RLC* circuit

- For $t > 0$: Applying KVL around the loop



$$\begin{cases} L \frac{di}{dt} + Ri + v = V_s \\ i = C \frac{dv}{dt} \end{cases} \rightarrow \frac{d^2v}{dt^2} + \frac{R}{L} \frac{dv}{dt} + \frac{v}{LC} = \frac{V_s}{LC}$$

$$v(t) = v_n(t) + v_f(t)$$

$v_n(t)$: natural response

$v_f(t)$: forced response

❖ Over damped: $v(t) = V_s + A_1 e^{s_1 t} + A_2 e^{s_2 t}$

❖ Critically damped: $v(t) = V_s + (A_1 + A_2 t) e^{-\alpha t}$

❖ Under damped: $v(t) = V_s + (A_1 \cos \omega_d t + A_2 \sin \omega_d t) e^{-\alpha t}$

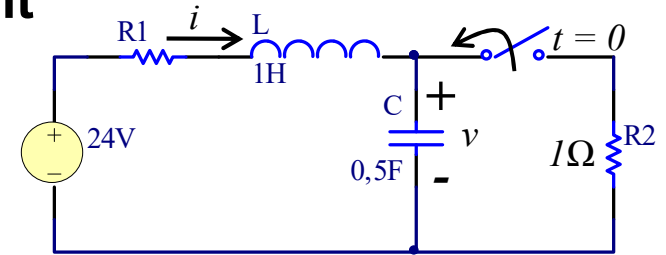
A_1, A_2 : determine from the initial conditions: $v(0)$,

$dv(0)/dt$

IV. Step response of a series/parallel *RLC* circuit

IV.1. Step response of a series *RLC* circuit

Ex 8.5: Find $v(t)$ and $i(t)$ for $t > 0$ in the case of the different values of $R_1 = 5\Omega, 4\Omega, 1\Omega$



➤ $R_1 = 5\Omega$

$$\begin{aligned} \text{❖ For } t = 0: i(0) = \frac{24}{R_1 + R_2} = 4A, \quad v(0) = 1 \cdot i(0) = 4V \rightarrow \begin{cases} \alpha = \frac{R_1}{2L} = 2,5 \\ \omega_0 = \frac{1}{\sqrt{LC}} = 2 \end{cases} \\ \rightarrow s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} \rightarrow \begin{cases} s_1 = -1 \\ s_2 = -4 \end{cases} \end{aligned}$$

$$\text{❖ For } t > 0: v(t) = v_f + A_1 e^{-t} + A_2 e^{-4t} = 24 + A_1 e^{-t} + A_2 e^{-4t}$$

$$\text{❖ At } t = 0: v(0) = 24 + A_1 + A_2 = 4 \rightarrow A_1 + A_2 = -20$$

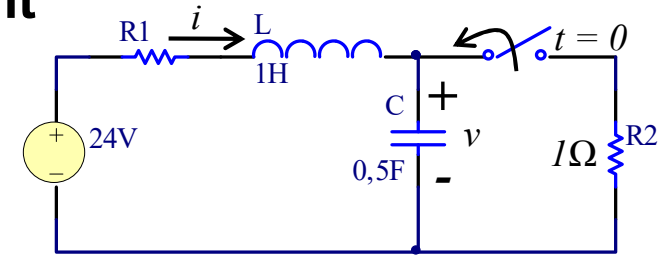
$$i(t) = C \frac{dv(t)}{dt} = C(-A_1 e^{-t} - 4A_2 e^{-4t}) \rightarrow i(0) = C(-A_1 - 4A_2) = 4$$

$$\rightarrow v(t) = 24 + \frac{4}{3}(-16e^{-t} + e^{-4t}) V \quad \rightarrow i(t) = C \frac{dv}{dt} = \frac{4}{3}(4e^{-t} - e^{-4t}) A$$

IV. Step response of a series/parallel *RLC* circuit

IV.1. Step response of a series *RLC* circuit

Ex 8.5: Find $v(t)$ and $i(t)$ for $t > 0$ in the case of the different values of $R_1 = 5\Omega, 4\Omega, 1\Omega$



➤ $R_1 = 4\Omega$

$$\text{❖ For } t = 0: i(0) = \frac{24}{R_1 + R_2} = 4,5A; v(0) = 1 \cdot i(0) = 4,5V \rightarrow \begin{cases} \alpha = \frac{R_1}{2L} = 2 \\ \omega_0 = \frac{1}{\sqrt{LC}} = 2 \end{cases}$$

$$\text{❖ For } t > 0: v(t) = v_f + (A_1 + A_2 t)e^{-2t} = 24 + (A_1 + A_2 t)e^{-2t}$$

$$\text{❖ At } t = 0: v(0) = 24 + A_1 = 4,5 \rightarrow A_1 = -19,5$$

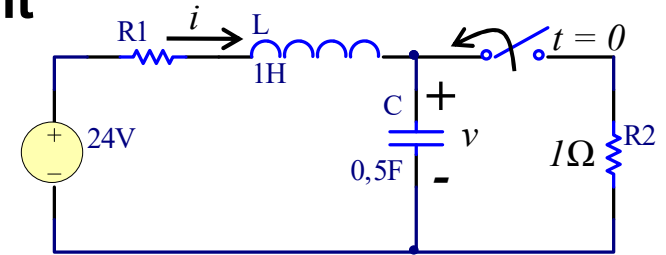
$$i(t) = C \frac{dv(t)}{dt} = C(-2A_1 - 2tA_2 + A_2)e^{-2t} \rightarrow i(0) = C(-2A_1 + A_2) = 4,5 \rightarrow A_2 = 57$$

$$\rightarrow v(t) = 24 + (-19,5 + 57t)e^{-2t} V \quad \rightarrow i(t) = C \frac{dv}{dt} = (4,5 - 28,5t)e^{-2t} A$$

IV. Step response of a series/parallel *RLC* circuit

IV.1. Step response of a series *RLC* circuit

Ex 8.5: Find $v(t)$ and $i(t)$ for $t > 0$ in the case of the different values of $R_1 = 5\Omega, 4\Omega, 1\Omega$



➤ $R_1 = 1\Omega$

$$\begin{aligned} \text{❖ For } t = 0: i(0) &= \frac{24}{R_1 + R_2} = 12A; v(0) = 1 \cdot i(0) = 12V \rightarrow \begin{cases} \alpha = \frac{R_1}{2L} = 0,5 \\ \omega_0 = \frac{1}{\sqrt{LC}} = 2 \end{cases} \\ &\rightarrow s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} = -0,5 \pm j1,936 \end{aligned}$$

$$\text{❖ For } t > 0: v(t) = 24 + (A_1 \cos 1,936t + A_2 \sin 1,936t)e^{-0,5t}$$

$$\text{❖ At } t = 0: v(0) = 24 + A_1 = 12 \rightarrow A_1 = -12$$

$$\frac{dv(t)}{dt} = e^{-0,5t} (-1,936A_1 \sin 1,936t + 1,936A_2 \cos 1,936t) - 0,5e^{-0,5t} (A_1 \cos 1,936t + A_2 \sin 1,936t)$$

$$\rightarrow \frac{dv(0)}{dt} = 1,936A_2 - 0,5A_1 = \frac{i(0)}{C} = 24 \rightarrow A_2 = 9,3$$

$$v(t) = 24 + (9,3 \sin \omega_0 t - 12 \cos \omega_0 t)e^{-0,5t} V \quad i(t) = (18,582 \sin \omega_0 t + 24 \cos \omega_0 t)e^{-0,5t} A$$

IV. Step response of a series/parallel *RLC* circuit

IV.2. Step response of a parallel *RLC* circuit

- Applying KCL at the top node for $t > 0$

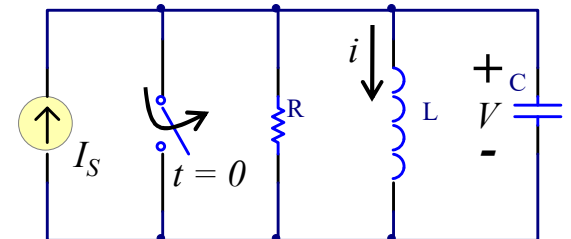
$$\frac{v}{R} + i + C \frac{dv}{dt} = I_s \xrightarrow{v=L \frac{di}{dt}} \frac{d^2 i}{dt^2} + \frac{1}{RC} \frac{di}{dt} + \frac{i}{LC} = \frac{I_s}{LC}$$

$$\rightarrow i(t) = i_n(t) + i_f(t) \quad \begin{array}{l} i_n(t): \text{natural response} \\ i_f(t): \text{forced response} \end{array}$$

- ❖ Over damped: $i(t) = I_s + A_1 e^{s_1 t} + A_2 e^{s_2 t}$

- ❖ Critically damped: $i(t) = I_s + (A_1 + A_2 t) e^{-\alpha t}$

- ❖ Under damped: $i(t) = I_s + (A_1 \cos \omega_d t + A_2 \sin \omega_d t) e^{-\alpha t}$



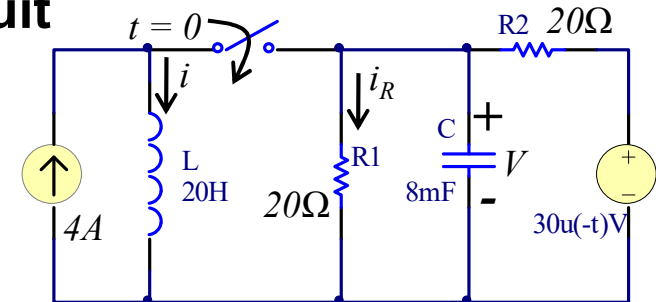
A_1, A_2 : determine from the initial conditions:

$i(0), di(0)/dt$

IV. Step response of a series/parallel RLC circuit

IV.2. Step response of a parallel RLC circuit

Ex 8.6: Find $i(t)$, $i_R(t)$ for $t > 0$



$$\text{For } t < 0: \begin{cases} i(0) = 4A \\ v(0) = \frac{R_1}{R_1 + R_2} 30 = 15V \rightarrow \frac{di(0)}{dt} = \frac{v(0)}{L} = 0,75 \end{cases}$$

$$\text{For } t > 0: \text{ We have a parallel RCL circuit with current source } R = \frac{R_1 R_2}{R_1 + R_2} = 10\Omega$$

$$\alpha = \frac{1}{2RC} = 6,25$$

$$\omega_0 = \frac{1}{\sqrt{LC}} = 2,5$$

$$\rightarrow s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} \rightarrow \begin{cases} s_1 = -11,978 \\ s_2 = -0,5218 \end{cases} \rightarrow i(t) = I_f + A_1 e^{-11,978t} + A_2 e^{-0,5218t}$$

$$\text{At } t = 0: \begin{cases} i(0) = 4 = 4 + A_1 + A_2 \rightarrow A_1 = -A_2 \\ \frac{di(0)}{dt} = -11,978A_1 - 0,5218A_2 = 0,75 \end{cases} \rightarrow \begin{cases} A_1 = -0.0655 \\ A_2 = 0.0655 \end{cases}$$

$$i(t) = 4 + 0.0655(e^{-0.5218t} - e^{-11.978t})A \quad i_R(t) = \frac{1}{20}L \frac{di}{dt} = 0,785e^{-11,978t} - 0,0342e^{-0,5218t}A$$



Chapter 8: Second-order circuits



V. General second order circuits

- Give a second order circuit, the step response $x(t)$ (current or voltage) can be determined by taking the following 5 steps:
 - ❖ Determine the **initial conditions** $x(0)$ and $dx(0)/dt$ and the **final value** $x(\infty)$
 - ❖ Find the **natural response** $x_n(t)$ (with 2 unknown constants) by turning off independent sources and applying KCL and KVL.
 - ❖ Obtain the **forced response** as: $x_f(t) = x(\infty)$
 - ❖ The total response is the sum of the natural response and forced response
$$x(t) = x_n(t) + x_f(t)$$
 - ❖ **Determine the 2 unknown constants** by imposing the initial conditions $x(0)$ and $dx(0)/dt$.

V. General second order circuits

Ex 8.7: Find the complete response v and i for $t > 0$.

- Find the initial and final values:

$$\begin{cases} v(-0) = 12V \\ i(-0) = 0 \end{cases} \rightarrow \begin{cases} v(+0) = v(-0) = 12V \\ i(+0) = i(-0) = 0 \end{cases}$$

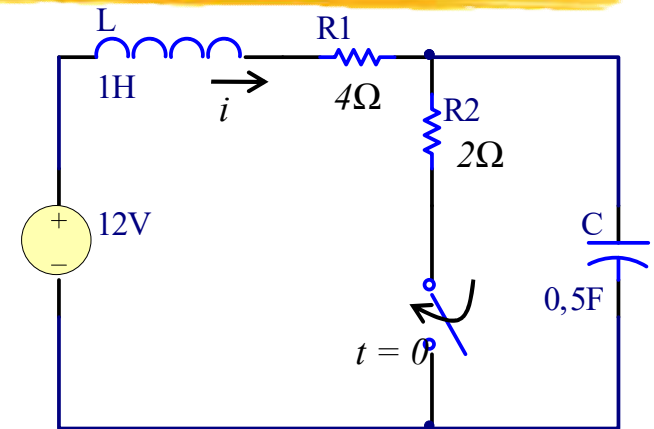
Applying KCL:

$$i(+0) = i_c(+0) + \frac{v(+0)}{R_2} \rightarrow i_c(+0) = -6A \rightarrow \frac{dv(+0)}{dt} = \frac{i_c(+0)}{C} = -12V/s$$

$$i(\infty) = \frac{12}{R_1 + R_2} = 2A, \quad v(\infty) = 2 \cdot i(\infty) = 4V = v_f(t)$$

- Find the natural response: Turn off the voltage source, and apply KCL, KVL

$$\begin{cases} i = \frac{v}{R_2} + C \frac{dv}{dt} \\ 4i + L \frac{di}{dt} + v = 0 \end{cases} \rightarrow \begin{cases} \frac{d^2v}{dt^2} + 5 \frac{dv}{dt} + 6v = 0 \\ s^2 + 5s + 6 = 0 \end{cases} \rightarrow v_n(t) = Ae^{-2t} + Be^{-3t}$$



V. General second order circuits

Ex 8.7: Find the complete response v and i for $t > 0$.

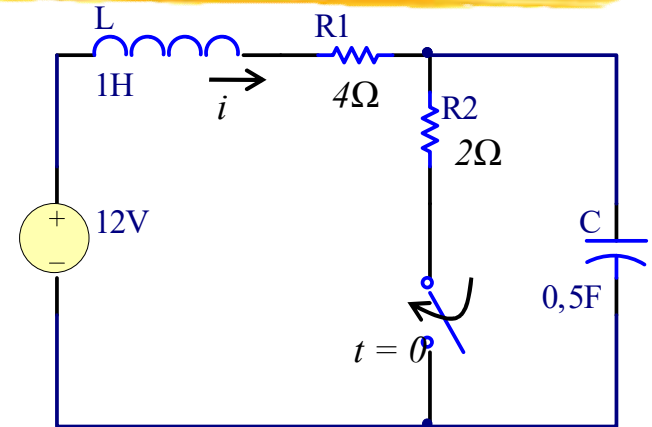
- The complete response is:

$$v(t) = v_n(t) + v_f(t) = 4 + A.e^{-2t} + B.e^{-3t}$$

- Imposing the initial condition gives:

$$\begin{cases} A + B = 8 \\ \frac{dv(0)}{dt} = -2A - 3B = -12 \end{cases} \rightarrow \begin{cases} A = 12 \\ B = -4 \end{cases} \rightarrow v(t) = 4 + 12e^{-2t} - 4e^{-3t} \text{ V}, \quad t > 0$$

$$i = \frac{v}{R_2} + C \frac{dv}{dt} = 2 + 6e^{-2t} - 2e^{-3t} - 12e^{-2t} + 6e^{-3t} = 2 - 6e^{-2t} + 4e^{-3t} \text{ A}, \quad t > 0$$



V. General second order circuits

Ex 8.8: Find $v_0(t)$ for $t > 0$

- Obtain the initial and final values of 2 currents

For $t < 0$: $7u(t) = 0 \rightarrow i_1(-0) = 0 = i_2(-0)$

$$\begin{aligned} i_1(+0) &= i_1(-0) = 0, \quad i_2(+0) = i_2(-0) = 0 \\ \rightarrow v_{L2}(+0) &= v_0(+0) = R_2 [i_1(+0) - i_2(+0)] = 0 \end{aligned}$$

- Applying KVL to the left loop at $t = +0$

$$7 = R_1 i_1(+0) + v_{L1}(+0) + v_0(+0) \rightarrow v_{L1}(+0) = 7V \rightarrow$$

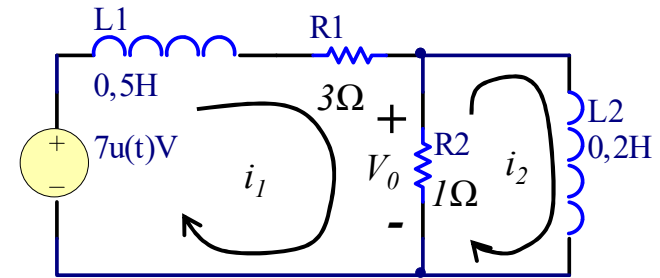
$$\frac{di_1(+0)}{dt} = \frac{v_{L1}}{L_1} = 14V/s$$

$$\frac{di_2(t)}{dt} = \frac{v_{L2}}{L_2} = 0$$

- As $t \rightarrow \infty$: $i_1(\infty) = i_2(\infty) = \frac{7}{R_1} = 2,33A$

- Applying KVL to 2 meshes to find natural responses:

$$\begin{cases} (R_1 + R_2)i_1 - R_2 i_2 + L_1 \frac{di_1}{dt} = 0 \\ R_2(i_2 - i_1) + L_2 \frac{di_2}{dt} = 0 \end{cases} \rightarrow \begin{cases} \frac{d^2 i_1}{dt^2} + 13 \frac{di_1}{dt} + 30 i_1 = 0 \\ s^2 + 13s + 30 = 0 \end{cases}$$



V. General second order circuits

Ex 8.8: Find $v_0(t)$ for $t > 0$

$$s^2 + 13s + 30 = 0 \rightarrow \begin{cases} s_1 = -3 \\ s_2 = -10 \end{cases} \rightarrow i_{1n} = Ae^{-3t} + Be^{-10t}$$

$$\rightarrow i_1(t) = i_{1f} + i_{1n} = 2,33 + Ae^{-3t} + Be^{-10t}$$

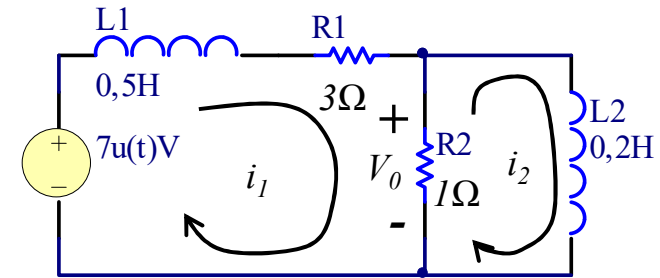
➤ Imposing the initial condition gives:

$$\begin{cases} A + B + 2,33 = 0 \\ -3A - 10B = 14 \end{cases} \rightarrow \begin{cases} A = -1,33 \\ B = -1 \end{cases} \rightarrow i_1(t) = 2,33 - 1,33e^{-3t} - e^{-10t} \text{ A}$$

➤ Applying KVL to loop

$$7 = 4i_1 - i_2 + L_1 \frac{di_1}{dt} \rightarrow i_2 = -7 + 4i_1 + L_1 \frac{di_1}{dt} = 2,33 - 3,33e^{-3t} + e^{-10t} \text{ A}$$

$$\rightarrow v_0(t) = R_2 [i_1(t) - i_2(t)] = 2(e^{-3t} - e^{-10t}) \text{ A}$$



V. General second order circuits

Ex 8.9: Find $v_o(t)$ for $t > 0$

➤ Applying KCL:

$$\text{At node 1: } \frac{V_s - V_1}{R_1} = C_2 \frac{dv_2}{dt} + \frac{V_1 - V_0}{R_2}$$

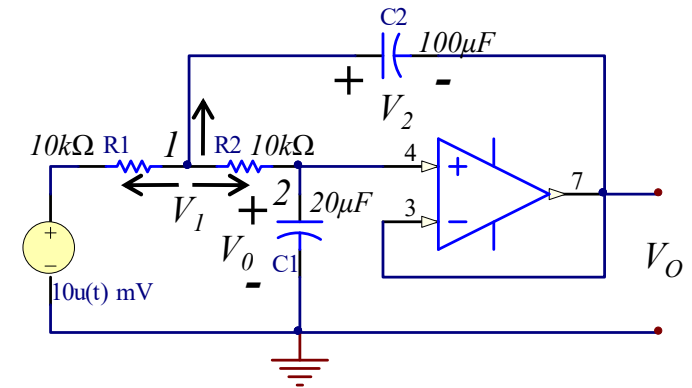
$$\text{At node 2: } \frac{V_1 - V_0}{R_2} = C_1 \frac{dv_0}{dt} \rightarrow V_1 = V_0 + R_2 C_1 \frac{dv_0}{dt}$$

$$\xrightarrow{V_2 = V_1 - V_0} \frac{V_s - V_1}{R_1} = C_2 \frac{dv_2}{dt} + C_1 \frac{dv_0}{dt}$$

$$\rightarrow \frac{d^2 v_0}{dt^2} + \left(\frac{1}{R_1 C_2} + \frac{1}{R_2 C_2} \right) \frac{dv_0}{dt} + \frac{V_0}{R_1 R_2 C_1 C_2} = \frac{V_s}{R_1 R_2 C_1 C_2} \rightarrow \frac{d^2 v_0}{dt^2} + 2 \frac{dv_0}{dt} + 5v_0 = 5v_s$$

➤ For the natural response, turning off the source: $s^2 + 2s + 5 = 0 \rightarrow s_{1,2} = -1 \pm j2$

$$\rightarrow v_{on}(t) = e^{-t} (A \cos 2t + B \sin 2t)$$



V. General second order circuits

Ex 8.9: Find $v_0(t)$ for $t > 0$

➤ As $t \rightarrow \infty$: $v_0(\infty) = v_1(\infty) = v_s \rightarrow v_{0f} = v_0(\infty) = 10\text{mV}$

➤ The complete response is:

$$\rightarrow v_0(t) = v_{0n} + v_{0f} = 10 + e^{-t} (A \cos 2t + B \sin 2t)$$

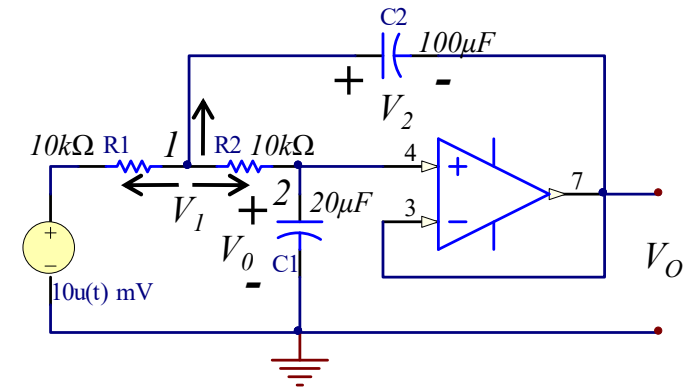
➤ Find initial conditions:

$$v_0(-0) = v_2(-0) = 0 \rightarrow \begin{aligned} v_0(+0) &= v_2(+0) = 0 \\ v_1(+0) &= v_2(+0) + v_0(+0) = 0 \end{aligned} \rightarrow \frac{dv_0(+0)}{dt} = \frac{v_1 - v_0}{R_2 C_1} = 0$$

$$\rightarrow \begin{cases} v_0(+0) = 10 + A = 0 \rightarrow A = -10 \\ \frac{dv_0(+0)}{dt} = -A + 2B = 0 \rightarrow B = -5 \end{cases}$$

➤ The complete response becomes:

$$\rightarrow v_0(t) = 10 - e^{-t} (10 \cos 2t + 5 \sin 2t) \text{ mV}$$





Chapter 8: Second-order circuits



VI. Applications

- Practical applications of RLC circuits are found in control and communications circuits, for examples:
 - ❖ Ringing circuits
 - ❖ Peaking circuits
 - ❖ Resonant circuits
 - ❖ *Smoothing circuits*
 - ❖ Filters
 - ❖ *Automobile ignition*
- Most of the circuits cannot be covered until we treat AC sources.